State-Dependent DMC with a Causal Helper

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Joint work with Amos Lapidoth

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State-dependent DMC

- Input, output, and state alphabets are \mathcal{X} , \mathcal{Y} , and \mathcal{S}
- State sequence is IID according to P_S
- Given input X = x and state S = s, output equals y with probability

W(y|x,s)

Capacity depends on whether or not channel-state information (CSI) is available, to whom, and how.

Causal CSI: Shannon's classic result

At time i, encoder knows s^i , so time-i input is produced via

$$x_i = f_i(m, s^i)$$

where m denotes the message.

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Theorem [Shannon '58]

Capacity with causal CSI at Transmitter is the maximum of

I(U;Y)

over joint distributions of the form

 $P_S(s)P_U(u)P_{X|US}(x|u,s)W(y|x,s).$

Without loss of optimality, x can be chosen as a deterministic function of (u, s).

How this capacity is achieved

- u is a mapping (or "strategy") that maps s to x
- \blacktriangleright Capacity is that of the "super channel" with input u and output y
- Notice that, at time *i*, the optimal encoder only uses s_i and ignores s^{i-1} :

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Related to the above: strictly causal CSI, where

$$x_i = f_i(m, s^{i-1}),$$

does not increase capacity.

This work: Imperfect causal CSI via Helper

At time *i*:

• Helper observes s^i and produces $t_i \in \mathcal{T}$:

 $t_i = h_i(s^i)$

(No additional constraint on Helper except that \mathcal{T} is fixed, with $|\mathcal{T}| < |\mathcal{S}|$)

• Transmitter sees t^i and produces input

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Main question: Is it optimal to choose $t_i = h_i(s_i)$ and $x_i = f_i(m, t_i)$?

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We shall also consider some variants of the above setting.

Some related works

- Rosenzweig, Steinberg, and Shamai '05:
 Full CSI at Receiver and quantized CSI at Transmitter
 (In the causal case, t_i = h_i(s_i) is assumed as part of the setup)
- Steinberg '08: Full CSI at Transmitter and quantized CSI at Receiver
- Bross, Lapidoth, and Marti '20 (two papers):
 Additive-noise channels with quantized CSI at Transmitter or at Receiver

Scalar-quantization lower bound

• Helper:
$$t_i = g(s_i)$$

Encoder: Shannon strategies treating t as the effective state: $x_i = u_i(t_i)$.

This achieves rate

with joint distribution

$$P_S(s)P_U(u)P_{T|S}(t|s)P_{X|UT}(x|u,t)W(y|x,s)$$

which satisfies three conditions:

$$\begin{array}{c} (U,T) & \longrightarrow & -(X,S) & \longrightarrow & -Y\\ S & \longrightarrow & -(U,T) & \longrightarrow & -X\\ U \perp & (S,T). \end{array}$$

(It's optimal to choose both $P_{T|S}$ and $P_{X|UT}$ to be deterministic.)

An upper bound

Define for every \boldsymbol{i}

$$U_i \triangleq (M, T^{i-1}, Y^{i-1}).$$

Then

$$n(R - \epsilon) \leq I(M; Y^n)$$

= $\sum_{i=1}^n I(M; Y_i | Y^{i-1})$
 $\leq \sum_{i=1}^n I(M, Y^{i-1}; Y_i)$
 $\leq \sum_{i=1}^n I(U_i; Y_i).$

 $\begin{array}{l} \operatorname{Recall} U_i \triangleq (M,T^{i-1},Y^{i-1}).\\ (U_i,T_i) = (M,T^i,Y^{i-1}) - & (X_i,S_i) - & -Y_i \quad \checkmark \end{array}$

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$$U_i = (M, T^{i-1}, Y^{i-1}) \rtimes (S_i, T_i) \quad \mathsf{X}$$

For example, both T and S binary, $T_{i-1}=S_{i-1}$ and $T_i=S_i\oplus S_{i-1}$

$$U_i = (M, T^{i-1}, Y^{i-1}) \rtimes (S_i, T_i) \quad \mathsf{X}$$

For example, both T and S binary, $T_{i-1}=S_{i-1}$ and $T_i=S_i\oplus S_{i-1}$

We only have

$$U_i \perp L S_i$$

i.e., joint distribution looks like

 $P_S(s)P_U(u)P_{T|US}(t|u,s)P_{X|UT}(x|u,t)W(y|x,s)$

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▶ If neither works, then we need to think about a better lower bound, but

Why on earth would Helper want to tell Transmitter anything about S^{i-1} at time i?

A variant (special case): S is part of the output

i.e. CSI is available to Receiver (causality plays no role).

(For clarity, Y denotes the part of output without S.)

Lower bound on capacity becomes

$$I(U;Y,S) = I(U;Y|S) = I(X;Y|S)$$

for joint distribution

$$P_S(s)P_{T|S}(t|s)P_{X|T}(x|t)W(y|x,s)$$

where it's again optimal to choose $P_{T|S}$ to be deterministic.

\boldsymbol{S} known to Receiver: Upper bound

Define

$$U_i \triangleq (M, Y^{i-1}), \qquad V_i \triangleq S^{i-1}$$

Then

$$\begin{split} n(R-\epsilon) &\leq I(M;Y^n,S^n) \\ &= \sum_{i=1}^n I(M;Y_i,S_i|Y^{i-1},S^{i-1}) \\ &\leq \sum_{i=1}^n I(M,Y^{i-1};Y_i,S_i|S^{i-1}) \\ &= \sum_{i=1}^n I(M,Y^{i-1};Y_i|S^{i-1},S_i) \\ &= \sum_{i=1}^n I(U_i;Y_i|V_i,S_i) \end{split}$$

S known to Receiver: Upper bound (contd.)

So capacity is upper-bounded by the maximum of

I(U; Y|V, S)

over distributions of the form

 $P_V(v)P_S(s)P_{U|V}(u|v)P_{T|SV}(t|s,v)P_{X|TUV}(x|t,u,v)W(y|x,s)$

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Observation: V can be removed via maximization over V = v, yielding

$$I(U;Y|S) = I(X;Y|S)$$

over

$$P_S(s)P_{T|S}(t|s)P_{X|T}(x|t)W(y|x,s)$$

which coincides with lower bound.

\boldsymbol{S} known to Receiver: Result

Theorem

When states are known to the Receiver, capacity is given by

 $\max I(X;Y|S)$

where \max is over joint distributions of the form

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Note: in this variant it is optimal to choose

$$t_i = h(s_i), \qquad x_i = f_i(m, t_i).$$

Another variant: Helper knows message ${\cal M}$

That is, at time i, Helper produces

$$t_i = h_i(\mathbf{m}, s^i)$$

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Our upper bound in the original setting continues to hold in this case:

capacity $\leq \max I(U;Y)$

over

$$P_S(s)P_U(u)P_{T|US}(t|u,s)P_{X|UT}(x|u,t)W(y|x,s).$$

Helper knows M: Achievability

• Codebook: $\{u^n(m), m \in \mathcal{M}\}$ all generated IID according to P_U .

• Helper knows m and hence also $u^n(m)$; it generates

 $t_i = h(u_i, s_i)$

Now t_i does depend on u_i !

Encoder generates

 $x_i = f(u_i, t_i)$

Helper knows M: Result

Theorem

When the message is known to the helper, capacity is given by

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Again, it is optimal to let t_i not depend on s^{i-1} !

Helper knows M: An example

State contains two independent uniform bits:

$$S = (S^{(0)}, S^{(1)})$$

Helper has one bit to use:

$$\mathcal{T} = \{0, 1\}$$

Input contains two bits

$$X = (A, B), \qquad A, B \in \{0, 1\}$$

Output is also two bits

 $Y = (A, B \oplus S^{(A)})$

Helper knows M example: Solution

We can send two information bits (k, ℓ) as follows:

- Help is $T = S^{(k)}$
- Transmitter sends $(k, \ell \oplus T)$
- Output is then (k, ℓ) (no decoding needed)

It is easy to prove that the above is optimal, so

capacity = 2 bits.

(We can also use the capacity formula to get this.)

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First consider scalar quantizer (independent of M).

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Given T = t, we have a "sum channel":

• Channel 1: $X = (0, B), Y = (0, B \oplus S^{(0)})$; capacity is $1 - H(S^{(0)}|T = t)$

▶ Channel 2: X = (1, B), $Y = (1, B \oplus S^{(1)})$; capacity is $1 - H(S^{(1)}|T = t)$

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Capacity of the sum channel is given by

$$\log\left(2^{1-H(S^{(0)}|T=t)} + 2^{1-H(S^{(1)}|T=t)}\right)$$

Same example when Helper does *not* know M (contd.)

$$\log \left(2^{1-H(S^{(0)}|T=t)} + 2^{1-H(S^{(1)}|T=t)} \right)$$

$$\leq \log \left(2 - H(S^{(0)}|T=t) + 2 - H(S^{(1)}|T=t) \right)$$

$$\leq \log \left(4 - H(S^{(0)}, S^{(1)}|T=t) \right)$$

Averaging over T and noting that \log is concave, we have that capacity—with a scalar-quantization Helper who doesn't know M—is at most

$$\log\left(4 - H(S^{(0)}, S^{(1)}|T)\right) \le \log 3$$

(This bound is tight: It can be achieved when Helper always sends $T = S^{(0)}$).

Example when Helper is non-scalar (does not know M) Allow Helper to be noncausal, and provide T to both transmitter and receiver. First consider $t = h(s_1, s_2)$. Given T = t, capacity of the "sum channel" is $\log\left(2^{2-H(S_1^{(0)},S_2^{(0)}|T=t)} + 2^{2-H(S_1^{(0)},S_2^{(1)}|T=t)} + 2^{2-H(S_1^{(1)},S_2^{(0)}|T=t)} + 2^{2-H(S_1^{(1)},S_2^{(1)}|T=t)}\right)$ $= \log \left(2^{1-H(S_1^{(0)}|T=t)} \cdot 2^{1-H(S_2^{(0)}|S_1^{(0)},T=t)} + 2^{1-H(S_1^{(0)}|T=t)} \cdot 2^{1-H(S_2^{(1)}|S_1^{(0)},T=t)} \right)$ $+ \ 2^{1-H(S_1^{(1)}|T=t)} \cdot 2^{1-H(S_2^{(0)}|S_1^{(1)},T=t)} + 2^{1-H(S_1^{(1)}|T=t)} \cdot 2^{1-H(S_2^{(1)}|S_1^{(1)},T=t)} \bigg)$ $\leq \log \left(2^{1-H(S_1^{(0)}|T=t)} \cdot 2^{1-H(S_2^{(0)}|S_1^{(0)},S_1^{(1)},T=t)} + 2^{1-H(S_1^{(0)}|T=t)} \cdot 2^{1-H(S_2^{(1)}|S_1^{(0)},S_1^{(1)},T=t)} \right)$ $+ 2^{1-H(S_1^{(1)}|T=t)} \cdot 2^{1-H(S_2^{(0)}|S_1^{(0)},S_1^{(1)},T=t)} + 2^{1-H(S_1^{(1)}|T=t)} \cdot 2^{1-H(S_2^{(1)}|S_1^{(0)},S_1^{(1)},T=t)} \Big)$ $= \log \left(2^{1-H(S_1^{(0)}|T=t)} + 2^{1-H(S_1^{(1)}|T=t)} \right) + \log \left(2^{1-H(S_2^{(0)}|S_1,T=t)} + 2^{1-H(S_2^{(1)}|S_1,T=t)} \right)$ $\leq \log\left(4 - H(S_1|T=t)\right) + \log\left(4 - H(S_2|S_1, T=t)\right) \leq 2\log\left(4 - \frac{1}{2}H(S_1, S_2|T=t)\right)$

Averaging over T we again get $\log 3$ per channel use.

The proof can be generalized to $t = h(s^n)$ for any finite n.

Example: Summary

Theorem

In the above example, when Helper knows the message,

capacity = 2 bits.

When Helper does not know the message,

capacity $= \log 3$.

Both equations hold irrespectively of whether Helper is causal or noncausal. Furthermore, they hold irrespectively of whether or not the help is also given to the Receiver.

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We have learned from this example:

- 1. Helper knowing the message ${\cal M}$ can indeed make a difference.
- 2. There is indeed a gap between our lower and upper bounds in the original setting (where Helper does not know M).

Let's go back to the original problem

Recall our upper bound was

$$\frac{1}{n}\sum_{i=1}^{n}I(U_i;Y_i)$$

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Roughly speaking, possible dependence between U_i and T_i comes via T^{i-1} .

We want to find a situation where it is useful to convey T^{i-1} to the Receiver.

WARNING

The following example may cause headache to some audiences.

Example where scalar-quantization Helper is suboptimal

- ▶ State is two uniform bits as in previous example: $S = (S^{(0)}, S^{(1)})$
- Input has three parts: X = (A, B, C) where A and B are binary, while |C| = 2^η
- Output also has three parts: $Y = (A', D^{(0)}, D^{(1)})$ where A' is binary, while $D^{(0)}$ and $D^{(1)}$ each contains η bits
- Helper has one bit: $T = \{0, 1\}$

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Channel law is:

If B ≠ S^(A), then Y ⊥⊥ (X, S) and uniform over its alphabet
 If B = S^(A), then

$$A' = A, \qquad D^{(B)} = C, \qquad D^{(B\oplus 1)} \perp X.$$

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You really want B to equal $S^{(A)}$...

Coding scheme for this example with non-scalar quantizer

Channel law:

▶ If
$$B \neq S^{(A)}$$
, then $Y \perp (X, S)$
▶ If $B = S^{(A)}$, then $A' = A$, $D^{(B)} = C$, $D^{(B \oplus 1)} \perp X$.

Let $T_0 \triangleq 0$. At time *i*,

$$\begin{split} T_i &= S_i^{(T_{i-1})} \\ A_i &= T_{i-1} \\ B_i &= T_i \\ C_i \text{ carries } \eta \text{ information bits} \end{split}$$

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How this scheme works:

- Always guaranteed that $B = S^{(A)}$
- At time *i*, Receiver learns B_{i-1} , so it can recover C_{i-1} from $D_i^{(B_{i-1})}$ \implies we achieve η bits per channel use

Example: Scalar-quantization Helper is always worse

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▶ If $B \neq S^{(A)}$, then $Y \perp (X, S)$ ▶ If $B = S^{(A)}$, then A' = A, $D^{(B)} = C$, $D^{(B \oplus 1)} \perp X$.

First consider $T_i = S_i^{(0)}$.

- We must avoid $B \neq S^{(A)}$, so we must always choose A = 0 and B = T.
- $D^{(0)}$ or $D^{(1)}$ equals C, but Receiver does not know which one.

For large $\eta,$ the achieved rate $\approx \eta-1$ bits.

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Other scalar quantizers are even worse. We can list all possible scalar quantizers and upper-bound the rate that is achievable with each quantizer. Proof omitted.

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▶ This can be generalized to a block-Markov scheme; details omitted.



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- Main finding: for a DMC with a memoryless state sequence, scalar-quantization Helper + Shannon strategy at Transmitter need not be optimal
- They become optimal if states are revealed to Receiver
- If Helper knows the message, then (message-dependent) scalar quantizer is optimal
- Helper knowing the message increases capacity

Thank you!